

# BOONTON

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#### SAFETY SUMMARY

The following general safety precautions must be observed during all phases of operation and maintenance of this instrument. Failure to comply with these precautions or with specific warnings elsewhere in this manual violates safety standards of design, manufacture, and intended use of the instrument. Boonton Electronics assumes no liability for the customer's failure to comply with these requirements.

# THE INSTRUMENT MUST BE GROUNDED

To minimize shock hazard the instrument chassis and cabinet must be connected to an electrical ground. The instrument is equipped with a three conductor, three prong a.c. power cable. The power cable must either be plugged into an approved three-contact electrical outlet or used with a three-contact to a two-contact adapter with the (green) grounding wire firmly connected to an electrical ground at the power outlet.

# DO NOT OPERATE THE INSTRUMENT IN AN EXPLOSIVE ATMOSPHERE.

Do not operate the instrument in the presence of flammable gases or fumes.

# KEEP AWAY FROM LIVE CIRCUITS.

Operating personnel must not remove instrument covers. Component replacement and internal adjustments must be made by quailfied maintenance personnel. Do not replace components with the power cable connected. Under certain conditions dangerous voltages may exist even though the power cable was removed, therefore; always disconnect power and discharge circuits before touching them.

# DO NOT SERVICE OR ADJUST ALONE.

Do not attempt internal service or adjustment unless another person, capable of rendering first aid and resuscitation, is present.

# DO NOT SUBSTITUTE PARTS OR MODIFY INSTRUMENT.

Do not install substitute parts or perform any unauthorized modification of the insturment. Return the instrument to Boonton Electronics for repair to ensure that the safety features are maintained.

#### SAFETY SYMBOLS.



This safety requirement symbol (located on the rear panel) has been adopted by the International Electrotechnical Commission, Document 66 (Central Office) 3, Paragraph 5.3, which directs that and instrument be so labeled if, for the correct use of the instrument, it is necessary to refer to the instruction manual. In this case it is recommended that reference be made to the instruction manual when connecting the instrument to the proper power source. Verify that the correct fuse is installed for the power available, and that the switch on the rear panel is set to the applicable operating voltage.



The CAUTION sign denotes a hazard. It calls attention to an operation procedure, practice, or the like, which, if not correctly performed or adhered to, could result in damage to or destruction of part or all of the equipment. Do not proceed beyond a CAUTION sign until the indicated conditions are fully understood and met.



The WARNING sign denotes a hazard. It calls attention to an operation procedure, practice, or the like, which, if not correctly performed or adhered to, could result in injury or loss of life. Do not proceed beyond a WARNING sign until the indicated conditions are fully understood and met.



Indicates dangerous voltages.

# TABLE OF CONTENTS

# Section | General information

Paragr	raph . P	age
1-1 1-2 1-4 1-7 1-9	Introduction	1-1
	Section II Installation	
2-1 2-3 2-5 2-7 2-10 2-11	Introduction	2-1 2-1 2-1 2-1
	Section !!! Operation	
3-1 3-3 3-5 3-7 3-8 3-10 3-11 3-13	Introduction	3-1 3-1 3-1 3-1 3-1
	Section IV Theory of Operation	
4-1 4-3 4-7 4-9 4-13 4-14 4-16 4-17 4-18 4-19 4-21 4-32	Introduction	4-1 4-2 4-3 4-3 4-3 4-4 4-4 4-4
	Section V Maintenance	
5-1 5-3 5-5 5-7 5-10 5-11 5-12 5-14 5-15 5-16 5-17 5-18	Introduction	5-1 5-1 5-1 5-1 5-1 5-1 5-1

# Section V Maintenance (Continued)

Parag	graph <sup>f</sup>	Page
5-20 5-22 5-24 5-25 5-26 5-27 5-28 5-29 5-30 5-31 5-36	Probe Frequency Response Test Swept Frequency Response and SWR test Adjustments Chopper Adjustments Delay Adjustment Zeroing Adjustment Output Integrator Adjustments Range Adjustments Troubleshooting Trouble localization	5-4 5-6 5-6 5-6 5-6 5-6 5-6
•	Section VI Parts List	
6-1	Introduction	6-1
	Section VII Schematic Diagrams	
7-1	Table of Contents	7-1
	List of Illustrations	
Figur	re 1	Page
1-2 2-1 3-1 3-2	Model 92EA RF Millivoltmeter	3-2 3-2 3-2 3-6 4-1 4-2 4-3 4-6 5-3 5-4
	List of Tables	
3-2 3-3 3-4 5-1 5-2 5-3 6-1	Specifications Operating Controls, Indicators and Connectors Measurements Rear Panel Pins: Unit Loadings and Comments Interface Input Characteristics Maintenance Test Equipment Power Supply Tests Points Troubleshooting Chart Manufacturer's Federal Supply Code Numbers List of Replaceable Parts	3-4 3-6 3-6 5-2 5-1 5-7 6-1

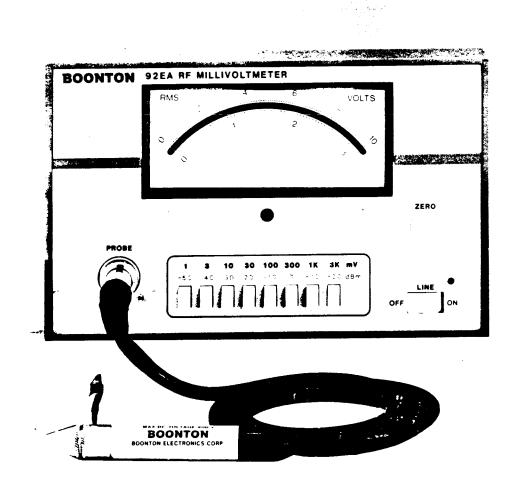


Figure 1-1. Model 92EA RF Millivoltmeter.

#### SECTION I GENERAL INFORMATION

### 1-1. INTRODUCTION

This instruction manual provides general information, installation, operation, theory of operation, maintenance instructions, parts list, and schematics for the Model 92EA RF Millivoltmeter.

#### 1-2. DESCRIPTION

The Model 92EA is a solid state, sensitive, accurate, and sturdily contructed RF Millivoltmeter that is characterized by high input impedance, excellent stablity, and low noise.

- 1-3. The 92EA design features are as follows:
- a. Frequency Range. Provides voltage measurements over the frequency range of 10 kHz to 1.2 GHz and 10 Hz to 100 MHz with the -16 Option. Relative accuracy above 1.2 GHz is typically  $\pm$  0.5 dB. Relative accuracy refers to the differential between two measured levels without regard to the absolute accuracy of either measurement. A measurement of a 100 mV signal at 8 GHz may indicate 20 mV on the meter; then a 50 mV signal, at that same frequency, will be indicated as 10 mV, within about 0.5 dB (5.9%).
- b. Voltage Range. The voltage range is 200  $\mu$ V to 3 V (300 V with the 100:1 divider accessory. The voltmeter has eight ranges, from 1 mV, fs, to 3 V, fs, arranged in a 1-3-10 sequence. The instrument exhibits true RMS response for input signals up to 30 millivolts gradually approaching peak-to-peak above this level. The meter; however, is calibrated to indicate RMS of a sine wave above 30 mV.
- c. Low Noise. The instrument is designed and constructed to hold noise from all sources to a minimum. The probe cable is of special low-noise design; a vigorous flexing causes only momentary, minor deflections of the meter on the most-sensitive range. The probe itself is insensitive to shock or to vibration.

Amplification takes place at 94 Hz, reducing susceptibility to any 50 or 60 Hz line-frequency-related fields. The input signals from the probe are converted into 94 Hz signals by a solid-state chopper.

- d. Minimal Zero Adjustment. Zero adjustment is not required on the upper five sensitivity ranges of the voltmeter. For measurements on the lower three ranges, the ZERO control is adjusted on the most sensitive range before operation. Only infrequent checking will be required during the course of subsequent measurements.
- e. DC Output. The instrument provides a linear DC output whose current capability of 1 mA into 1000 ohms is extremely stable. When used as part of an automatic test

system, the fast response of the instrument's DC output to an input step-function allows many tests per unit time.

#### 1-4. ACCESSORIES

- 1-5. The following accessories are supplied with the instrument:
- a. 952001 RF Probe. Probe with low-noise cable and connector assembly for measurements from 10 kHz to 1.2 GHz.
- c. 952002 50  $\Omega$  BNC Adapter. Used for measurements up to 1.2 GHz in a 50  $\Omega$  system.
- b. 952004 Probe Tip. Removable probe tip with grounding-clip lead; for use up to approximately 100 MHz.
- 1-6. The following accessories are optional and may be ordered:
- a. 91-6F Unterminated Type F Adapter (F).
- b. 91-6G Unterminated BNC Adapter (M).
- c. 91-8B/1A 75  $\Omega$  Type F Adapter (F).
- d. 91-14A 50  $\Omega$  Tee Adapter N (F/F). Permits connection into 50  $\Omega$  line; operational to 1.2 GHz.
- e. 91-15A 50  $\Omega$  Termination N (M).
- f. 91-16A Unterminated N Adapter. May be used with all probes. Used for coaxial connection up to approximately 100 MHz, or to 400 MHz when fed from a 50 ohm source in an electrically short system.
- g. 950000 Rack Mtg. Kit, Single. Mounts one unit left or right of blank panel in 19 inch rack. 5.25 inches high.
- h. 950001 Rack Mtg. Kit, Dual. Mounts two units side by side in 19 inch rack. 5.25 inches high.
- i. 950002 Single Rack-Mounting Kit. Kit for mounting one 92EA as one-half of a module in a standard 19-inch rack.
- j. 950031 Transit Case.
- k. 952003 50  $\Omega$  Tee Adapter. Type-N Tee connector used with 952014 termination, it permits connection into a 50 ohm line.
- I. 952005 100:1 Voltage Divider. Attenuates input signal by a factor of  $100 \pm (1 + f_{\rm MHz}/200)$  %, permitting measurements up to 300 V, and extending the rms measuring range to 3 V; also increases input resistance by a factor of 1000 to 3000, depending upon input level. Operates from 50 kHz to 700 MHz. Maximum input potential, 1000 V, DC plus peak AC.
- m. 952006 75  $\Omega$  BNC Adapter. Used for measurements up to 500 MHz in a 75 ohm system.

#### SECTION I GENERAL INFORMATION

- n. 952007 75  $\Omega$  Tee Adapter. Type-N Tee connector; used with 952015 termination it permits connection into a 75 ohm line.
- o. 952008 Unterminated BNC Adapter (Female). Used for coaxial connection up to approximately 100 MHz, or to 400 MHz when fed from a 50 ohm source in an electrically short system.
- p. 952011 Accessory Kit, 50  $\Omega_{\rm e}$  Kit contains the following items: 952008, unterminated BNC adapter (F); 952005, 100:1 voltage divider; 952003, 50  $\Omega$  Tee adapter; 952014, 50  $\Omega$  termination; Model 952103, Storage Case (case for protecting and storing kit accessories).
- q. 952012 Accessory Kit, 75  $\Omega_{\bullet}$  Contains the following items; 952008, unterminated BNC adapter; 952005, 100:1 divider; 952007, 75 ohm Tee adapter; 952015, type-N 75  $\Omega$  termination; and 952013 storage case.
- r. 952013 Accessory Case. For use with the model 952001 probe and accessories.
- s. 952016-1 Low Frequency Probe. 10 Hz to 100 MHz. Overload protection, 10 VAC and 50 VDC.
- t. 952058 100:1 Divider. For use with 952016 Low Frequency Probe; frequency range 10 Hz to 20 MHz.

#### 1-7. OPTIONS

- 1-8. The following options are available:
- a. -04 dBV Scale Uppermost. The dBv scale is uppermost on the meter face. The two voltage scales are above and below the mirror.
- b. -06 75  $\Omega$  dBm Scale Uppermost. The dBm scale is uppermost on the meter face. The 952006 75  $\Omega$  BNC adapter is supplied instead of the standard 952002 50  $\Omega$  BNC adapter.
- c. -08 Rear Signal Input. An additional RF probe input connector is located on the rear panel.
- d. -12 dBmV Scale Uppermost. The dBV scale is uppermost on the meter face. The 952006 75  $\Omega$  BNC adapter is supplied instead of the standard 952002 50  $\Omega$  BNC adapter.
- e. -16 Low Frequency Version. Frequency Range 10 Hz to 100 MHz, utilizes the 952016-1 probe.

#### 1-9. SPECIFICATIONS

1-10. Specifications are listed in Table 1-1.

#### TABLE 1-1. SPECIFICATIONS

Voltage Range: 200  $\mu\text{V}$  to 3 V (300 V up to 700 MHz with accessory 100:1 voltage divider). Lowest detectable voltage is approximately 100  $\mu\text{V}_{\bullet}$ 

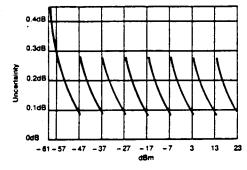
Full-Scale Voltage Ranges: 1, 3, 10, 30, 100, 300, 1000, and 3000 mV.

dBm Range: -61 to +23 dBm (with the 50 ohm adapters. -63 to +21 dBm (with the 75 ohm adapters.

Frequency Range: 10 kHz to 1.2 GHz (uncalibrated response to approximately 8 GHz).

Accuracy: The maximum uncertainty is the sum of the uncertainties given in sections A, B, and  $C_{\star}$ 

- A1. Basic Uncertainty, Voltage Ranges: 200  $\mu\text{V}$  3000 mV, 1% fs.
- A2. Basic Uncertainty, dBm:



#### B1. Frequency Effect: (952001 Probe):

50  $\Omega$  measurements using the 952001 Probe with the 952002 BNC Adapter or terminated 952003 Type-N Tee Adapter at 100 mV level.

Frequency	mV	dBm
1 MHz (cal freq)	0	0
100 kHz - 100 MHz	1% rdq	0.09 dB
100 MHz - 1 GHz	3% rdq	0.27 dB
1 GHz - 1.2 GHz	7% rda	0.63 dB

# B2. Frequency Effect: (952016 Probe) -16 Option:

50  $\Omega$  measurements using the 952016 Probe with the 952002 BNC Adapter.

Frequency	mV	dBm
1 MHz (cal freq)	0 rdq	0
50 Hz - 20 MHz	1% rolg	0.09 dB
20 Hz - 50 Hz	2% rdq	0.17 dB
10 Hz - 100 MHz	5% rolg	0.44 dB

# B3. Frequency Effect: (952016 Probe) with 952058 100:1 Divider:

Frequency	mV	dBm
1 MHz (cal freq)	0	0
1 MHz - 20 MHz	5% rdg	0.44 dB
50 Hz - 1 MHz	3.5% rda	0.31 dB
20 Hz - 50 Hz	4.5% rdq	0.40 dB
10 Hz - 20 Hz	7.5% rdg	0.68 dB
SWR: 1.05 to 100 MHz.	7.5 A TOG	0.000

# TABLE 1-1. SPECIFICATIONS (Continued)

#### C. Temperature Effect, at 1 MHz:

Temperature	mV Ra	nges
Range	Instrument	RF Probe
21°C to 25°C 18°C to 30°C 10°C to 40°C 0°C to 55°C	0 0.2% rdg. 0.5% rdg. 1% rdg.	0 1% rdq. 5% rdq. 12.5% rdg.

Temperature	dBm Ra	inges
Range	Instrument	RF Probe
21°C to 25°C 18°C to 30°C 10°C to 40°C 0°C to 55°C	0 0.02 dB 0.04 dB 0.09 dB	0 0.09 dB 0.45 dB 1.16 dB

Maximum AC input: 10 V, All frequencies and ranges.

Maximum DC input: 400 V, All ranges.

Meter: 4<sup>1</sup>/<sub>2</sub>-inch taut-band. Two linear voltage scales:

0 to 3; resolution 0.05/division. 0 to 10; resolution 0.1/division.

OW. zogarithmic dBm scale:
-10 to +3; resolution 0.2/division, max

Meter Unrest: (1 mV fs range, only).

Indicated Voltage	Unrest
Above 600 цV	<1 <b>%</b> fs
300 μV to 600 μV	<2% fs
200 uv to 300 uv	<5% fs

EMI: There is no significant radiated or conducted leakage from the instrument or the probe.

Waveform Response: True RMS response for input levels up to 30 mV (3 volts to 700 MHz using the 100:1 Voltage Divider), with transition to peak-to-peak (calibrated in RMS) at higher levels.

#### Crest Factor:

Direct input:	Level	300 µV	1 mV	3 mV	10 mV	30 mV
	C.F.	140	42	14	4.2	1.4
With Divider:	Level	30 mV	100 mV	300 mV	1 V	3 V
	C.F.	140	42	14	4.2	1.4

SWR: Less than 1.15 to 1.2 GHz (return Loss greater than 23 dB).

DC Output: 0 to 10 VDC, proportional to RF input voltage. Source resistance of 9 k $\Omega$ ; will deliver 1 mA into 1 k $\Omega$  load. Full-scale input step function response time less than 100 ms on 30 mV, fs, to 3 V, fs, ranges, increasing to 1 s on the 1 mV, fs, range.

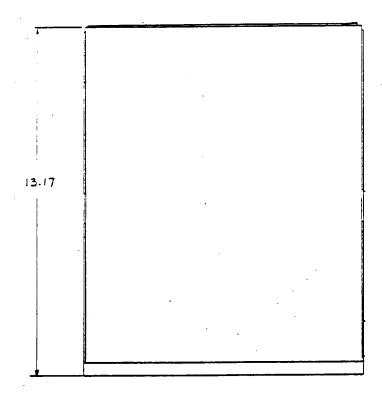
Power: 100, 120, 220, 240 V  $\pm$ 10%, 50 to 400 Hz.

Operating and Storage Temperatures:

Operating: 0°C to +55°C Storage: -55°C to +75°C

Dimensions: 5.83 in (14.9 cm) high  $\times$  8.34 in (21.1 cm) wide and 13.17 in (33.3 cm) deep. Refer to Figure 1-2 Outline Dimensions.

Weight: Net 3.2 kg (7 lbs).



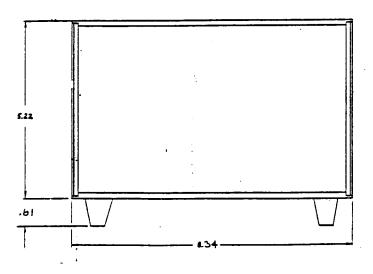


Figure 1-2. Outline Dimensions.

# SECTION II

#### 2-1. INTRODUCTION

2-2. This section contains the unpacking, mounting, power requirements, line voltage selection, cable connections and the preliminary checkout procedure.

#### 2-3. UNPACKING

2-4. The instrument is shipped complete and is ready to use upon receipt. Unpack the instrument from its shipping container and inspect it for damage that may have occurred during shipment. Refer to Figure 2-1.

#### NOTE

Save the packing material and container for possible use in reshipment of the instrument.

#### 2-5. MOUNTING

2-6. For bench mounting, choose a clean, sturdy, uncluttered mounting surface. For rack mounting, an accessory kit is available.

#### 2-7. POWER REQUIREMENTS

2-8. The instrument has a tapped power transformer and two line voltage selection switches which permit operation from 100, 120, 220 and 240 volt  $\pm 10\%$ , 50 to 60 Hz, single phase AC power sources. Power consumption is approximately 5 VA.

#### CAUTION

Always make certain that the line voltage selector switches are set to the correct positions most nearly corresponding to the voltage of the available AC power source, and that a fuse of the correct rating is installed in the fuse holder before connecting the instrument to any AC power source.

2-9. Set the line voltage selector switches, located on the rear panel to the appropriate positions as indicated on the LINE VOLTAGE SELECT chart located next to the switches. Check that the line fuse is correct for the selected power source.

VOLTAGE	FUSE
100/120 V	0.2 A MDL (SB)
220/240 V	0.1 A MDL (SB)

#### 2-10. CABLE CONNECTIONS

- 2-11. Cable connections required depend on the use of the instrument. Cable connections that may be required are as follows:
- a. Connect the probe cable to the front panel PROBE jack or if Option -08 is installed connect the probe cable to the rear panel PROBE jack.
- b. An ANALOG OUTPUT connector is located on the rear panel which provides a connection to monitor a DC voltage proportional to the meter reading.
- c. A REMOTE connector is located on the rear panel which provides a connection for operation with a remote interface.

#### 2-11. PRELIMINARY CHECKOUT PROCEDURE

- a. Make sure that the serial number of the probe to be used is the same as that of the voltmeter. (Each instrument is calibrated for its particular RF probe.) Use of a probe other than that for which the instrument was calibrated may result in measurement errors.
- $\ensuremath{\text{b}}_{\bullet}$  Connect the probe cable to the PROBE jack on the front panel.
- c. Connect the power cord to the instrument and the desired power source. Refer to paragraph 2-7 for proper power application.
- d. Set the LINE ON/OFF power switch to ON.
- e. Press the 1 mV range pushbutton; the panel meter pointer should rest on zero. If it does not, use the ZERO control to set the meter to zero. (This adjustment will hold for the other ranges.) The instrument is now ready for use.

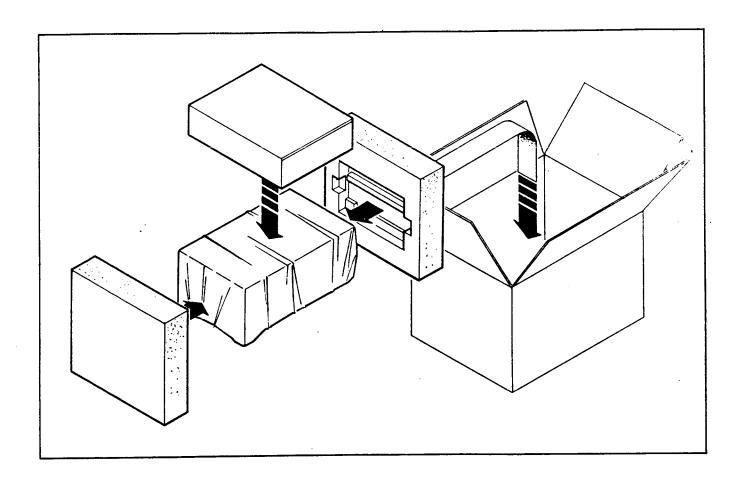


Figure 2-1. Packing and Unpacking Diagram.

# SECTION III OPERATION

#### 3-1. INTRODUCTION

3-2. Section III contains the operating controls, indicators, and connectors descriptions and functions, and operating instructions for the Model 92EA.

# 3-3. OPERATING CONTROLS, INDICATORS, AND CONNECTORS

3-4. The controls, indicators, and connectors used during operation of the instrument are listed in Table 3-1 and shown in Figures 3-1 and 3-2.

#### 3-5. OPERATING INSTRUCTIONS

- **3-6.** The operating instructions for the Model 92EA are as follows:
- a. Inital Conditions.
- b. Measurements.
- c. Interface Operation.
- 3-7. Initial Conditions.
- a. Make sure that the serial number of the probe to be used is the same as that of the voltmeter. (Each instrument is calibrated for its particular RF probe.) Use of a probe other than that for which the instrument was calibrated may result in measurement errors.
- b. Connect the probe cable to the PROBE jack on the front panel.
- c. Connect the power cord to the instrument and the desired power source. Refer to paragraph 2-7 for proper power application.
- d. Set the LINE ON/OFF power switch to ON.
- e. Press the 1 mV range pushbutton; the panel meter pointer should rest on zero. If it does not, use the ZERO control to set the meter to zero. (This adjustment will hold for the other ranges.)
- f. Press the range pushbutton that is closest to the level of the measurement to be made. The instrument is now ready for use.

#### 3-8. Measurements.

3-9. The instrument is capable of making different types of measurements: Refer to the Table 3-2 for Measurement and Connection Requirements.

#### 3-10. Measurement considerations.

a. Temperature Effects. Over the range of 21°C to 25°C (70°F to 77°F), temperature effects for the instrument and the Model 950001 RF Probe are sensibly zero. Outside of these limits, inaccuracies can be expected

as described in the Specifications section. However, no permanent change in probe characteristics will result from exposure to any reasonable high or low temperature.

Inaccuracies of measurement resulting from temperature effects may occur shortly after soldering close to the probe tip, or when measurement are made close to heat sources such as resistors, heat sinks, vacuum tubes,

When making low level measurements (below approximately 2 millivolts) it is important to make sure that the probe has attained a uniform temperature throughout its body. A temperature gradient between the inside and the outside of the probe can generate a small thermal voltage that may add to the DC output of the detector diodes.

- b. Hum, Noise and Spurious Pick-up. When measuring low level signals, precautions should always be taken to avoid the possibility of errors of measurement resulting from hum, noise or stray RF pick-up.
  Although all low frequency hum and noise are attenuated at the input, it is still possible for unwanted high level signals to cause errors. In some cases it may be necessary to provide extra shielding around the probe connections to reduce stray pick-up. Typical sources of spurious radiation are: induction or dielectric heating units, diathermy machines, local radio transmitters, and grid dip meters.
- c. Signal Overload on 1 mV Range. On the most sensitive (1 mV) range, the application of a large AC signal overloads the amplifier and a short time is required for the high impedance input circuit to discharge. This effect is significant for signals above approximately 100 millivolts. Typically, application of a 1 volt signal will require a recovery time of about thirty seconds before subsequent measurements should be made on the 1 mV range.
- d. Correction Curves for Models 952003-01A AND 952007-01A. Refer to Figure 3-3 and use the correction curves to make corrections for transmission loss when using the Type N 50  $\Omega$  or 75  $\Omega$  Tee Adapters.

#### 3-11. Interface Operation.

3-12. Remote programming is accomplished by simultaneously shorting to common the Manual Disable and the appropriate range lines on the card edge connector located at the rear of the instrument. (In effect, when Manual Disable is brought to common, the front panel switches are disconnected). Refer to Figure 3-3 and Table 3-3 for a pictorial and tabular presentation of the rear panel programming inputs and data output connections.

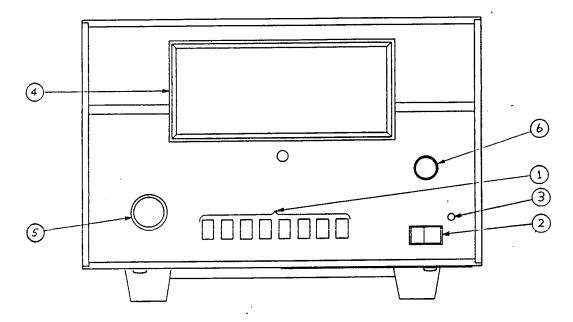


Figure 3-1. Model 92EA Front View.

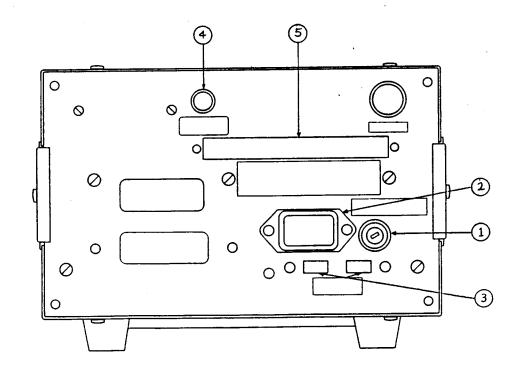


Figure 3-2. Model 92EA Rear View.

Control, Indicator,	Figure and Index No.	Function
FRONT PANEL		
FULL SCALE pushbuttons	3-1, 1	
mV dBm 3k +20 1k +10 300 0 100 -10 30 -20 10 -30 3 -40 1 -50		Depressing any full scale range pushbutton selects the operating range.
LINE ON/OFF switch	3-1,2	Turns the AC power ON or OFF.
LED indicator	3=1,3	Red light emitting diode is lit when the instrument is turned ON.
Meter	3-1,4	Taut-band meter with two linear voltage scales and one logarathmic dBm scale.
PROBE jack	3-1,5	Provides a connection for the probe cable.
ZERO control	3-1,6	Zeros the instrument on the three most sensitive ranges.
REAR PANEL		
Fuse holder	3-2,1	Fuse holder. 0.2A, for 100, 120 V; 0.1 A, for 220, 240 V.
AC connector	3-2,2	AC power connector.
Slide switches	3-2,3	Selects the proper operating voltage.
Analog Output	3-2,4	A DC voltage proportional to the meter reading is available at these terminals.
Remote connections	3-2,5	The card edge plug is intended for use with an Amphenol 225-2221-101 connector, or equivalent. See Table 3-3 for pin designations.

### 3-13. Input characteristics.

a. Interface input characteristics are given in Table 3-4.

b. Input Pull-Up. All input terminals have internal pull-up. The current sourced by this pull-up when the input is brought to a logic level 0 is included in the loading shown in the "Unit Loading" column of Table 3-4.

c. DC Analog Output.

<sup>1.</sup> Polarity: positive with respect to instrument ground. (The negative DC Analog Output terminal is at ground potential.)

<sup>2.</sup> Source Resistance: 9 k $\Omega_{\bullet}$ 

Table 3-	2. Measurements.
Measurement	Connection Requirements
Standard 92EA	
Frequency: 10 kHz - 1.2 GHz Voltage: 200 µV - 3 V Impedance: 50 Ohms	952001 RF Probe 952002 50 ohm BNC Adapter 952004 Probe Tip
Frequency: 10 kHz - 1.2 GHz Voltage: 3 V - 300 V Impedance: 50 Ohms	952001 RF Probe 952002 50 ohm BNC Adapter 952004 Probe Tip 952005 100:1 Divider
In Line Measurements	952003 50 ohm Tee Adapter (N) (F/F) (Optional)
92EA-06 and 92EA-12 Options	
Frequency: 10 kHz - 1.2 GHz Voltage: 200 µV - 3 V Impedance: 75 Ohms	952001 RF Probe 952006 75 ohm BNC Adapter 952004 Probe Tip
Frequency: 10 kHz - 1.2 GHz Voltage: 3 V - 300 V Impedance: 75 Ohms	952001 RF Probe 952006 75 ohm BNC Adapter 952004 Probe Tip 952005 100:1 Divider
In Line Measurement	952007 75 ohm Tee Adapter (N) (Optional)
92EA-16	
Frequency: 10 Hz - 100 MHz Voltage: 200 µV - 3 V Impedance: 50 Ohms	952016-1 RF Probe 952002 50 ohm BNC Adapter 952004 Probe Tip
Frequency: 10 Hz - 100 MHz Voltage: 3 V - 300 V Impedance: 50 Ohms	952016-1 RF Probe 952002 50 ohm BNC Adapter 952004 Probe Tip 952058 100:1 Divider
In Line Measurements	952003 50 ohm Tee Adapter (N) (F/F) (Optional)

#### CAUTION

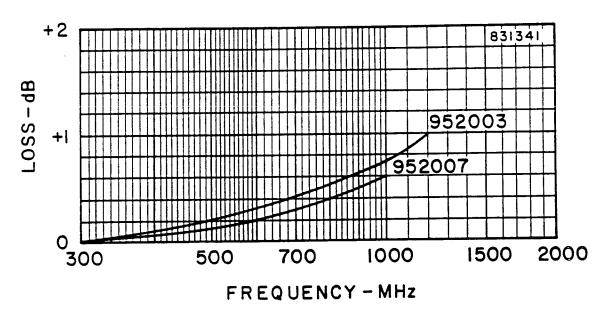
Overload Limits. The RF Probe supplied with the instrument is overload protected to 10 VAC, and to 400 VDC, EXCEEDING THESE LIMITS MAY RESULT IN PERMANENT DAMAGE TO THE PROBE.

The 952002 50 ohm adapter should not be subjected to continuous overload of more than 3 volts [DC + (AC, RMS)], to avoid excessive heating of the terminating resistor. Where voltages above these limits are likely to be encountered, the 100:1 Voltage Divider is required.

#### NOTE

When making measurements below 100 MHz, utilize the RF Probe supplied with the instrument. When making Measurements above 100 MHz, the RF probe tip should not be used. Connection should be made directly to the probe's center contact, with the ground connection kept as short as possible.

# CORRECTION FOR INSERTION LOSS



#### NOTES

- 1) The Insertion Loss shown is that which exists between the input and output ports of the Tee.
- 2) The RF Millivoltmeter measures the input voltage of the Tee.  $\,$
- 3) Therefore, if the output voltage of the Tee is to be determined, subtract the Insertion Loss determined from the graph from the value that is indicated on the instrument.
- 4) Do not use the correction if terminated measurements are required (i.e., measurements with Model 952028-01A or Model 952029-01A terminations installed on the output port of the Tee).

Figure 3-3. Correction Curves for Type N Tee Adapters Models 952003-01A (50  $\Omega$ ), and 952007-01A (75  $\Omega$ ).

	Table 3-3. Rear Panel Pins: Unit Loadings and Comments						
Pin No.	Function	Comment	Command	Unit Loading			
7	Man.Disable	Disables front-panel range selection	0	0.1			
16 15 14 13 12 11 10 9	1 mV range 3 mV " 10 mV " 30 mV " 100 mV " 300 mV " 1 V " 3 V "	Selects range, provided that Manual Disable has also been selected. Selecting more than one range will result in incorrect indications. Range lines must be de-selected for manual operation.	0000000	0.1 0.1 0.1 0.1 0.1 0.1 0.1			
5	Common DC Analog	Power line ground at rear panel, +10 V for full scale of "1" ranges; +9.5 V for "3" ranges.	N/A N/A	N/A N/A			

Tab	Table 3-4. Interface input Characteristics							
TTL Series	Logic Level	Voltage Level	Curent per Unit Load					
Standard Power 54/74	0	±0.7 V	±1.6 mA* 40 μA					

\*The "-" current indicates current out of the input (the external command device must sink this current). A standard power (Series 54/74) TTL output will sink and source 10 unit loads.

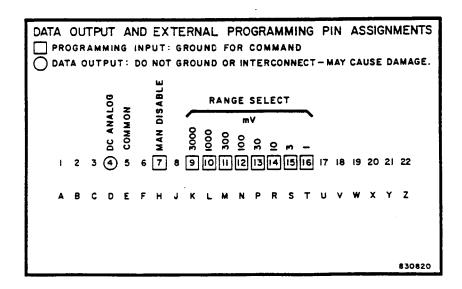


Figure 3-4. Rear Panel Pin Assignments.

# SECTION IV THEORY OF OPERATION

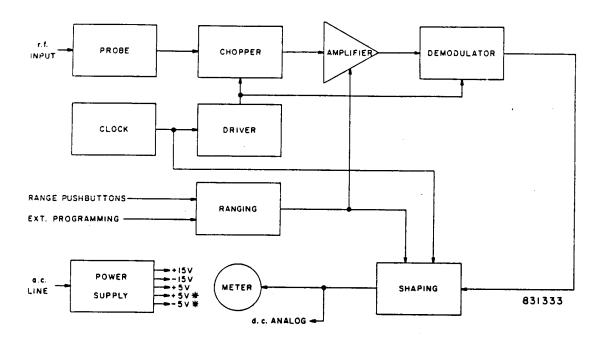


Figure 4-1. Functional Block Diagram.

#### 4-1. INTRODUCTION

4-2. The Model 92EA is a solid state, sensitive, accurate, and sturdily contructed RF Millivoltmeter that is characterized by high input impedance, excellent stability, and low noise.

# 4-3. FUNCTIONAL BLOCK DIAGRAM DESCRIPTION (Rafer to Figure 4-1.)

4-4. The RF voltage to be measured by the instrument is applied to a sensor, which converts the RF voltage to a proportional DC voltage. The output voltage from the sensor ranges from a fraction of a millivolt to volts, as a function of the input voltage level to the sensor. To reduce the effects of drift and residuals at very low levels, the DC output voltage of the sensor is applied to a solid state chopper, which converts the DC voltage to a 94 Hz square wave with an amplitude proportional to the DC voltage. The drive signals for the chopper are provided from the analog section.

- 4-5. The analog section provides amplification, ranging, and demodulation of the 94-Hz square-wave signal supplied from the chopper. Ranging is performed manually by means of eight pushbuttons on the front panel; remote ranging is available by means of rear panel programming connectors. The analog section also receives a 752 Hz clock signal from the clock section; the chopper and demodulator drive signals are derived from this clock signal by frequency divider circuits in the analog section. The DC output voltage of the analog section is supplied to the shaping section, where the amplified and demodulated DC voltage is converted to a linear voltage used for driving the meter and the analog output. The probe's output is inherently non-linear before shaping; it is true RMS on the lower ranges, and peak-to-peak on the higher ranges.
- **4-6.** Operating power for the instrument circuits is provided by the power supply. Line voltages of 100, 120, 220, or 240 volts,  $\pm 10\%$ , may be applied to the power transformer. Switches on the rear panel of the instrument allow the switching of primary

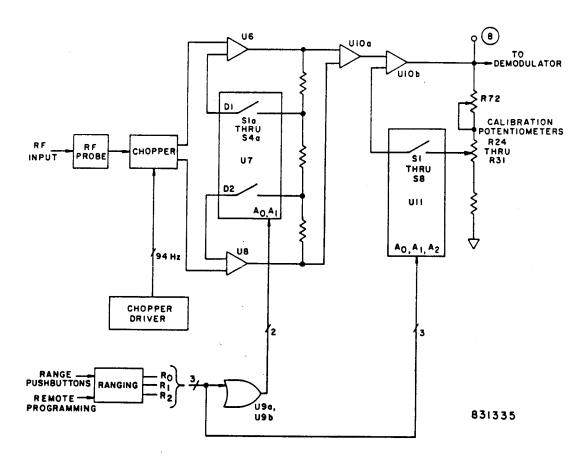


Figure 4-2. Amplifier Circuits Block Diagram.

### 4-6. (Continued).

winding connections in order to accomodate the various input voltages. The secondary windings of the power transformer furnish power to rectifier regulator circuits that develop regulated +5, -5, +15, and -15 volts for operation of the other circuits in the instrument.

#### 4-7. DETAILED CIRCUIT DESCRIPTIONS

4-8. In the following paragraphs the circuits of the instrument are grouped by function as follows: Sensor, Analog (includes Chopper, Amplifier, Demodulator and Driver), Clock, Ranging, Shaping, and Power Supply.

4-9. SENSOR. The RF probes used with the instruments are unterminated, high input-impedance devices. The RF voltage applied to the probes is rectified by a full-wave detector employing factory selected diodes with special characteristics, including low capacitance and controlled thermal offsets. The resulting DC voltage is a function of the applied RF voltage.

4-10. A full-wave detector, unlike a single diode detector, permits measurement of highly asymmetrical waveforms without substantial error. When a voltage of such waveform is

impressed on a single diode detector circuit, whether or not the portion of the waveform that "turns on" the diode is restricted to the square law region of the diode's characteristic, the recovered DC voltage is dependent upon the phase of the input voltage. Consider a positive pulse of low duty cycle applied to the anode of a diode detector. The recovered DC voitage is a function of the polarity, amplitude, and duration of the If the pulse is inverted, it drives the diode into a reverse bias condition and the recovered DC voltage is near zero. A full-wave detector circuit, on the other hand, yields an equal amount of DC irrespective of the polarity of the input pulse. The RMS voltage of a pulse obviously does not depend upon the phase of the pulse. The response of a single diode detector, square law characteristic notwithstand-ing, cannot reflect this; a full-wave detector circuit does.

4-11. Probe response is true RMS for inputs below 30 mV. Above this voltage level, the probe response gradually changes, approaching peak-to-peak at the higher levels. However, the voltage data is digitally shaped in the instrument to indicate RMS voltage, provided that the input is reasonably sinusoidal, as is the case with CW and FM signals.

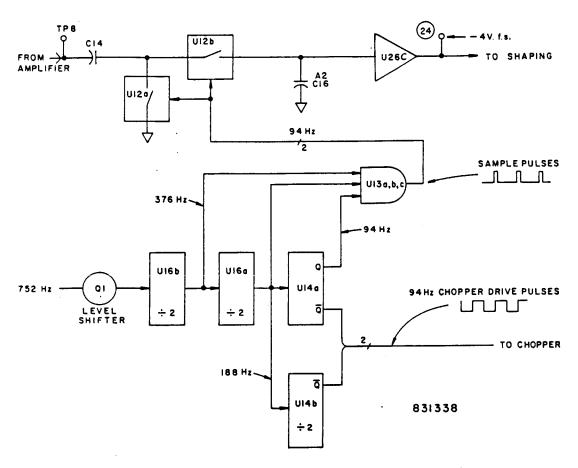


Figure 4-3. Demodulator and Driver Circuits Block Diagram.

**4-12.** The probe body has been designed to minimize noise. The probe connects to the instrument through a low noise cable.

4-13. CHOPPER CIRCUITS. The chopper board contains four solid state switches, which are used to convert input DC voltage to a 94 Hz square wave. The switches are controlled by 94-Hz chopper drive signals supplied from a frequency divider chain in the analog section. Potentiometers A4R4 and A4R5 provide means for adjusting the chopper to zero output with zero input. Use of a solid state chopper eliminates most of the undesirable characteristics of electromechanical choppers (e.g.: contact wear, bounce, and contamination). The output of the chopper is a balanced 94 Hz square wave that is directly proportional to the DC voltage applied from the sensor.

4-14. AMPLIFIER CIRCUITS. Refer to Figure 4-2. The balanced 94 Hz square wave signal from the chopper is amplified by operational amplifiers A2U6, A2U8, A2U10a and A2U10b. The gain of the operational amplifiers A2U6, A2U8 and A2U10a is controlled by adjusting feedback to the amplifier through multiplexer A2U7 and a resistor network. The signals R0, R1 and R2, from the ranging section, applied through gates A2U9a and A2U9b, control the

switching of input terminals D1 and D2 of multiplexer A2U7 to two of eight points in the resistance networks, thereby adjusting the feedback and the amplifier gain.

4-15. The 94 Hz output of op amps A2U6 and A2U8 is applied to the differential inputs of op amp A2U10a, which makes the signal single ended. This signal is amplified by op amp A2U10b and associated circuitry. Multiplexer A2U11 adjusts the gain of this op amp in eight steps, under control of signals RO, R1 and R2 from the ranging section, to provided decade ranging in voltage. The nominal output for a full scale input on each range is four volts, approximately, peak-to-peak (at TPB). Separate potentiometers are provided for full scale calibration of the instrument on each range. A2R72 is the Master Gain Control and A2R73 is used to adjust for any large differences in the effectency of probes.

4-16. DEMODULATOR CIRCUITS. Refer to Figure 4-3. A solid state demodulator, consisting of switches A2U12a and A2U12b, converts the amplified and scaled 94 Hz square wave signal back to DC. The demodulator is driven by a 94 Hz demodulator drive signal, which is synchronized with the 94 Hz chopper drive signal. A synchronous, sampling type demodu-

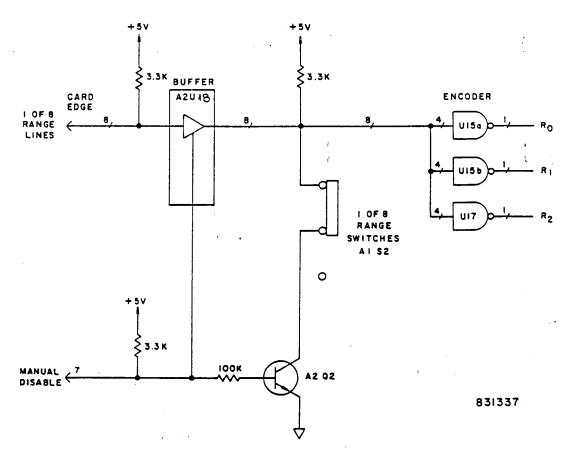


Figure 4-4. Ranging Circuits Block Diagram.

#### 4-16. (Continued).

lator circuit is used, with the sample being taken at a point well removed from the chopper switching points. The demodulator is followed by the high input impedance buffer A2U26c to reduce loading of the sampling capacitor A2C16 to negligible proportions. Output DC is suplied to A2U24c in the shaping section.

4-17. DRIVER CIRCUITS. Chopper and demodulator drive signals are derived from a 752 Hz signal supplied from the clock circuit. A2Q1 shifts the clock voltage from the zero to plus five volt level (used by the shaping circuitry), to a plus five and minus five volt level used by the chopper circuitry. Flip-flops A2U14a, A2U14b, A2U16a and A2U16b divide down the 752 Hz signal to 94 Hz, and gates A2U13a, A2U13b and A2U13c shape the 94 Hz demodulator signal. Figure 4-3 shows the derivation of the chopper and demodulator drive signals from the 752 Hz clock signal.

4-18. CLOCK CIRCUITS. Refer to Figure 4-5. A crystal controlled oscillator (A2U19a, b, c, and crystal Y1), provides 96 kHz clock pulses. This 96 kHz square wave is frequency divided in the binary counter A2U20 to produce clock signals for both the analog and the shaping circuitry.

4-19. RANGING AND PROGRAMMING CIRCUITS. Refer to Figure 4-4. Ranging of the instrument is performed manually by means of the eight pushbutton switches on the front panel. When a range is selected, one section of AIS2 "shorts" to ground inputs to the range encoder. The range encoder comprises A2U15a, A2U15b, and A2U17, and will generate a binary range code on lines RO, R1 and R2. This range code is used by both the amplifier circuitry and the shaping circuitry to select the appropriate signal processing.

4-20. Remote programming is accomplished by simultaneously shorting to common the Manual Disable and the appropriate range lines on the card edge connector located at the rear of the instrument. Shorting the Manual Disable line turns off A2Q2, causing the latter to disconnect the front panel switches. Shorting the Manual Disable line also enables buffer A2U18, allowing it to transmit ranging information from the card edge connector to the range encoder.

4-21. SHAPING CIRCUITS. Shaping is used to linearize the output of the diode sensor. (The conversion of RF to DC in the sensor is virtually square law for the lowest ranges, gradually becoming quasi-linear at three volts.) The shaping circuit of the instru-

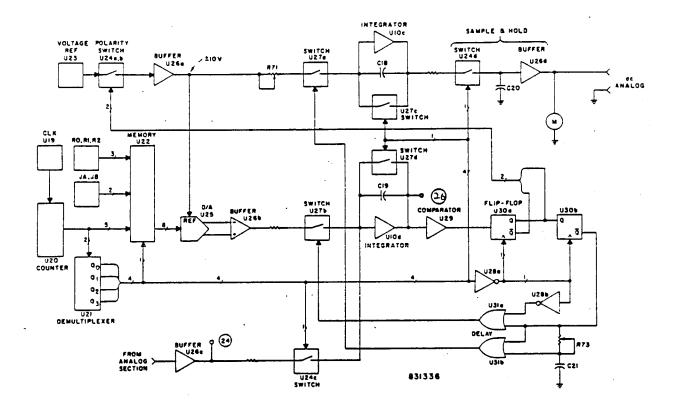


Figure 4-5. Shaping Circuits Block Diagram.

#### 4-21. (Continued).

ment uses data stored in ROM A2U22, and a dual slope integrator, to linearize the output of the sensor.

4-22. Shaping is achieved in four phases. Phase 0 integrates the input signal. Phase 1 combines the input with the shaping data. Phase 2 samples and holds the shaped and corrected output. Phase 3 resets the circuit for another cycle. The phases are controlled by the output of demultiplexer A2U21. Refer to Figure 4-5.

4-23. Phase 0 occurs when  $\overline{00}$  of the demultiplexer A2U21 is low. During this phase, the DC voltage from the demodulator is applied to the integrator A2U10d through switch A2U24c. The integrator output starts at zero before ramping to its final value. (A voltage of -4 at TP24 will cause the integrator output voltage at TP26 to be approximately +5 volts at the end of Phase 0.)

4-24. Comparator A2U29 measures the polarity of the integrator at TP26 and gives an output of either 0 V or 5 V for a negative or a positive integration, respectively. The polarity information is stored in filp-flop

A2U30a and is used to determine the polarity of the voltage reference at pin 1 of A2U26a. A2U23 is a stable voltage source. A2U24a and A2U24b are switches that configure amplifier A2U26a as either non-inverting or inverting, thus changing from (+) to (-) the polarity of the voltage derived from the reference.

4-25. Phase 1 occurs when  $\overline{\text{Q1}}$  of the A2U21 demultiplexer is low. This signal is inverted in A2U28a to provide a clock to the flip-flops A2U30a and A2U30b. The clock latches the polarity information into A2U30a and also sets Q2 of A2U30b low.

4-26. The Phase 1 signal addresses ROM A2U22. The ROM is addressed also by RO, R1 and R2 (the range lines, which set shaping appropriate to the range currently in use), and by JA and JB (the shaping jumpers, which program for variations in sensor shaping). ROM A2U22 also receives signals in a binary sequence from counter A2U20. Shaping data stored in the ROM are recalled and converted in DAC A2U25 into the shaping signal. The shaping signal is of the opposite polarity to the input signal. Closing switch A2U27b allows the shaping signal and the input signal to combine in the integrator. (A2U27b is closed by the Phase 1 signal moregated with the output of A2U30b.)

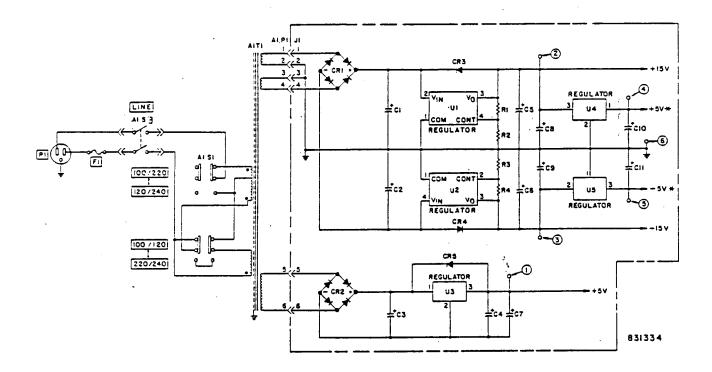


Figure 4-6. Power Supply Circuits Block Diagram.

- 4-27. As just noted, the shaping signal and the input signal are of opposite polarity. The integrator therefore ramps toward zero volts at a rate proportional to the shaping signal's amplitude.
- 4-28. Phase 1 closes switch A2U27a, allowing the reference voltage to be integrated in A2U10c, the output integrator. The output integrator provides a shaped signal that drives the meter and the DC analog output by way of A2U26d (a high impedance unity gain buffer). The time constant of A2R69 & A2C21 delays the closing of switch A2U27a in order to mask the noise caused by the shaping process when there is zero input. A2R71 adjust for the tolerance of the integrator capacitor.
- 4-29. During Phase 1 the comparator A2U29 detects the zero crossing of the integrator. When the zero crossing is detected, the output integrator switch A2U27a is opened, halting the integrator A2U10c at a voltage proportional to the sum of the shaping and the input voltages. The output of the comparator, and the output of the comparator flip-flop, are "exclusive or-ed" in A2U28b. This signal resets flip-flop A2U30b, opening both switches A2U27a and A2U27b.

- 4-30. During Phase 2, the Q2 output of demultiplexer A2U21 closes switch A2U24d. This samples the output of integrator A2U10c and holds the voltage in A2C20.
- 4-31. The  $\overline{03}$  output of demultiplexer A2U21 closes switches A2U27c and A2U27d during Phase 3. These switches discharge their respective integrator capacitors, leaving the integrators ready for another shaping cycle.
- 4-32. POWER SUPPLY CIRCUITS. Refer to Figure 4-6. The power supply circuits provide DC operating power for all other circuits of the instrument. Regulated output voltages of +15, -15, +5, and -5 volts are supplied. Line voltages of 100, 120, 220 and 240 volts, ±10%, 50 to 400 Hz, can be accommodated.
- 4-33. AC power is applied to the primary windings of power transformer AlT1 through the LINE switch AlS3, and the two section line voltage switch AlS1. The latter changes the transformer primary winding connections as required to accommodate the available line voltage. Fuse AlF1 protects the power-supply circuits against overload. The voltages developed in the secondary windings of the power transformer are applied to three rectifier regulator circuits on the main printed circuit board A2.

- 4-34. The +15 V and -15 V supplies are similar. Input to each supply consists of 20 V, supplied by a separate secondary winding of the power transformer. In each supply, the applied AC is rectified by the bridge rectifier A2CR1, filtered by A2C1 and A2C2, and then regulated by A2U1 and A2U2.
- 4-35. Regulated +5 V and -5 V operating supplies for the chopper and analog circuits derive power from the regulated +15 V and the -15 V supplies, using regulators A2U4 and A2U5. Thus, the supplies for these more sensitive circuits are isolated from the less sensitive circuits.
- **4-36.** Regulated +5 V for the digital circuits is provided by a separate 5 VDC supply, powered by the third secondary winding of the power transformer. This further isolates the sensitive input circuitry from noise caused by the digital circuitry. The applied AC is rectified by the bridge rectifier A2CR2 to develop 11 VDC, at A2C3. This filtered DC is converted to +5 V by regulator A2U3.

#### SECTION V MAINTENANCE

### 5-1. INTRODUCTION

5-2. This section contains the safety requirements, test equipment required, cleaning procedures, removal and replacement procedures, inspection procedures, adjustment procedures (calibration), and troubleshooting for the Model 92EA.

#### 5-3. SAFETY REQUIREMENTS

5-4. Although this instrument has been designed in accordance with international safety standards, general safety precautions must be observed during all phases of operation and maintenance of the instrument. Failure to comply with the precautions listed in the Safety Summary located in the front of this manual could result in serious injury or death. Service and adjustments should be performed only by qualified service personnel.

#### 5-5. TEST EQUIPMENT REQUIRED

5-6. Test equipment required for the performance tests, adjustments, and troubleshooting is listed in Table 5-1.

# 5-7. CLEANING PROCEDURE

5-8. Cleaning. Painted surfaces can be cleaned with a comercial, spray-type window cleaner or with a mild soap and water solution.

#### CAUTION

Avoid the use of chemical cleaning agents which might damage the plastics used in the instrument. The recommended cleaning agents is a solution of 1% mild detergent and 99% water.

#### 5-9. REMOVAL AND REPLACEMENT PROCEDURES

- 5-10. Instrument Covers. Remove the instrument covers as follows:
- a. Disconnect the power cord and all cables from the instrument.
- b. Remove the two screws located at the rear of the cover.
- c. Slowly lift the cover up and to the rear.
- d. Turn the instrument over and remove the bottom cover in the same manner as the top cover was removed.
- e. To replace the covers reverse the removal procedure.
- 5-11. RF Probe Replacement. If it is necessary to change probes for any reason, the instrument's calibration MUST be rechecked. In most cases, the instrument must be recalibrated. If the RF Probe supplied with the instrument is exchanged for a Low Frequency Probe, recalibration is required.

# 5-12. INSPECTION PROCEDURE

5-13. If an equipment malfunction occurs, perform a visual inspection of the instrument and RF probe. Inspect for signs of damage caused by excessive shock, vibration, or overheating, such as broken wires, loose hardware and parts, loose electrical connections, electrical shorts, cold solder connections, or accumulations of dirt and other foreign matter. Correct any problems discovered, then perform the performance tests to varify that the instrument is operational. If a maifunction persist or the instrument falls any of the performance tests, refer to the adjustment procedures. After the has been adjusted, perform the performance tests again to verify instrument operation. If the instrument can not be adjusted, or fails the performance tests refer to the troubleshooting procedure.

# 5-14. PERFORMANCE TESTS

5-15. Power Supply Tests. Improper operation of the instrument may be caused by incorrect DC operating voltages. Perform the power supply tests as follows:

#### WARNING

Line voltages up to 240 volts, AC may be encountered in the power supply circuits. To protect against electrical shock, observe suitable precautions when connecting and disconnecting test equipment, and when making voltage measurements.

- a. Refer to paragraph 5-9 Removal and Replacement Procedures and remove the instrument covers.
- a. Use voltmeter and measure the voltage at the test points listed in Table 5-2.

Supply Tests Points.
Voltage
+5.0 ±5% VDC
+15.0 ±5% VDC
-5.0 ±5% VDC
+5.0 ±5% VDC

- 5-16. Instrument Calibration Test. Check the calibration of the instrument on each range using a test voltage equal to the full scale value.
- 5-17. Probe Tests. If a probe exhibits out of tolerance performance in these tests, the user is urged not to attempt repair but to send the probe back to the factory for repair or replacement.

	TABLE 5-1. MAINTENANCE TEST EQUIPMENT					
EQUIPMENT	SPECIFICATIONS	SUGGESTED MODEL				
Signal Generator	125 kHz - 175 MHz 450 kHz - 1040 MHz 10 MHz - 1400 MHz	Boonton Model 1030 Boonton Model 102F-20 HP 8660 A/C				
Slotted line	50 ohm, 300 Mhz - 9 GHz	GR Type 900 LB				
Detector	.1 MHz - 1200 MHz 40 MHz - 2030 MHz	Boonton Model 9200A GR Type 1241				
Power Splitter	Type N, DC to 18 GHz Type N, DC to 18 GHz Type N, DC to 2000 MHz	HP 11667A Weinschel Model 1870A MicroLab/FXR DA-4FN				
RF Power Meter	.1 MHz - 18 GHz, 1.2% acc. .1 MHz - 26.5 GHz, 1.2% acc.	Boonton 4200A HP 435B				
Sweep Generator	Frequency range; 0.01 to 2600 MHz. 10 kHz - 2600 MHz	Wavetek 2001 HP 8660 A/C				
SWR Autotester	50 ohm, 10 MHz - 4 GHz	Wiltron Model 63NF50				
Oscilloscope	100 MHz, 2 channel	Tektronix Model 4658				
Standard Mismatch	Male, Type N 1.2:1 VSWR DC to 18 GHz	Maury Microwave Model 2562C				
DVM .	3 1/2 digit, .1≸ accuracy	Data Precision Model 1350				

#### 5-18. Probe SWR Test.

- a. Test equipment required: Refer to Table 5-1.
- 1. Signal Generator.
- 2. Slotted line.
- 3. Detector.

#### 5-19. Perform the test as follows:

- a. Connect the slotted line to a proper signal source, and terminate the line with the device to be tested, i.e., Boonton Model 952003-01A Tee Adapter and Boonton Model 952028-01A 50  $\Omega$  Termination, or Boonton Model 952002-01A 50  $\Omega$  Adapter.
- b. The probe and RF Millivoltmeter must be connected to the accessory being tested. The probe supplies a perturbation for which the accessory was designed, and which it needs, to meet its specification. The millivoltmeter permits the test level to be set to the desired value.
- c. Move the carriage of the slotted line to a point of minimum voltage, then to a point of maximum voltage, and record the values.

- d. The SWR is the ratio of the maximum and the minimum voltages. The measurement can be repeated at other frequencies and levels, as required.
- 5-20. Probe Frequency Response Test.
- a. Test equipment required: Refer to Table 5-1.
- 1. Sweep Generator.
- 2. Power Splitter.
- . 3. Calibrated RF Power Meter.
- 5-21. Perform the test as follows:
- a. Connect the equipment as shown in Figure 5+1.
- b. Set the frequency of the generator to 10 MHz, and adjust the output control for the desired test level. If the response is to be measured at one level only, a test voltage of 100 or 200 mV is recommended.
- c. Disable the output of the generator momentarily and zero the power meter. Reestablish the output level and note the reading on the power meter. Record the fre-

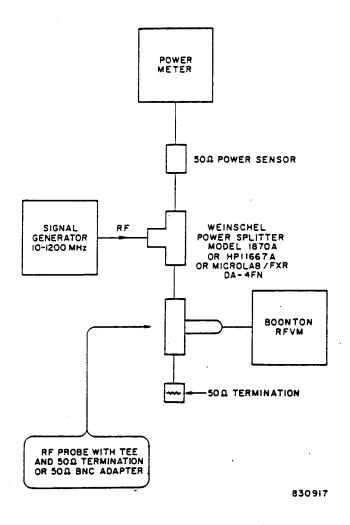


Figure 5-1. Frequency Response Test Setup.

#### 5-21. (Continued).

quency of the generator and the reading of the instrument. Change the frequency, in whatever increments are desired, through the range of 10 to 1200 MHz, holding the reference reading on the power meter constant.

- d. Reverse the output ports of the power splitter and repeat Step c.
- e. The correct voltmeter reading is obtained at each frequency by averaging the two readings. This virtually eliminates the influence of frequency differences of the two ports of the power splitter.
- f. Further refinements can be made by filtering the output of the generator, and measuring the complex reflection coefficients of the power meter, RF millivoltmeter accessory under test, and all ports of the power splitter. The usual corrections can then be made. These procedures are not

usually necessary, and should be done only if the additional accuracy is warranted.

#### 5-22. Swept Frequency Response and SWR Test.

- a. Test equipment required: Refer to Table 5-1.
- 1. Sweep Generator.
- 2. SWR Autotester.
- 3. Oscilloscope.
- 4. Power Splitter.
- 5. Standard 1.2:1 Mismatch Termination.

#### 5-23. Perform the test as follows:

a. Connect the equipment as shown in Figure 5-2, and temporarily connect the probe under test to the instrument. Adjust the output control of the sweep generator for a reading on the instrument of 100 mV at a fixed frequency of 100 MHz.

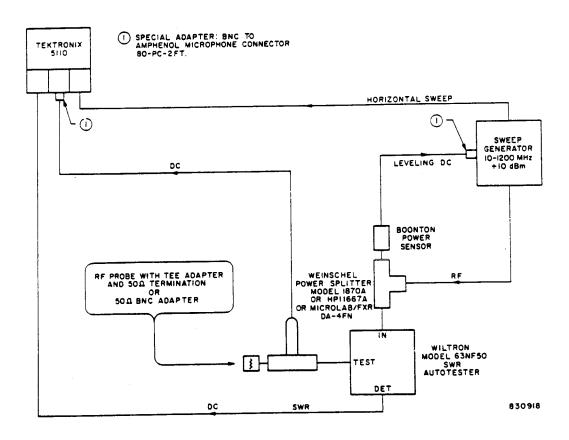


Figure 5-2. Swept Frequency Response Test Setup.

#### 5-23. (Continued).

b. Calibrate one of the scope's vertical-input amplifiers for a sensitivity of 100  $\mu\text{V/div}$ . The other vertical amplifier should be calibrated so that a change from 100 mV to 90 mV applied to the input of the probe under test will produce a vertical deflection of two divisions. This can be done easily if a Boonton Model 26A RF Millivoltmeter Calibrator is available. The probe should be temporarily connected to the output of the calibrator while output levels of 100 mV and 90 mV are alternately selected, and the sensitivity of the second input amplifier is adjusted for a deflection of two divisions.

- c. Substitute the Standard 1.2:1 Mismatch Termination for the accessory under test, and calibrate the graticule of the oscilloscope for an SWR of 1.2. Replace the accessory and probe.
- d. Adjust the limits on the three bands of the sweep generator for coverage from 10 to 1200 MHz. Study the traces for both frequency response and SWR (return loss).
- e. Reverse the output ports of the power splitter and repeat Steps c and  $\mbox{\bf d}_{\bullet}$
- f. Note that the permissible error for the frequency-response trace expands with frequency. For meaningful results, the grati-

cule should be marked with a grease pencil, showing maximum permissible errors for the various frequency bands as determined with a calibrated signal of, say, 1 MHz, and at levels above and below the selected test level. Note also that the recovered DC from the RF probe, which is applied to the second vertical amplifier, will vary as the square of the RF input level for test levels of 30 mV or less. Above 30 mV, the RF to DC conversion gradually changes from square law to linear, and approaches a peak-to-peak rectifier at an input of 3 volts.

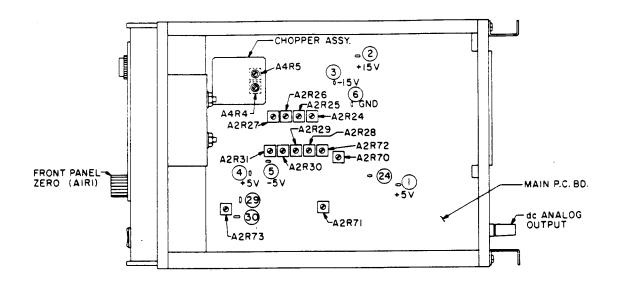
# 5-24. ADJUSTMENTS Refer to Figure 5-3.

5-25. Chopper Adjustments. Perform the adjustments as follows:

#### NOTE

The instrument must be turned on for not less than one hour and the instrument and its sensor must be at an ambient temperature of  $21^{\circ}\text{C}$  to  $25^{\circ}\text{C}$ .

- a. Refer to paragraph 5-9 Removal and Replacement Procedures and remove the top instrument cover.
- b. Remove the instrument "zero" potentiometer connector, A1P1.



831340

ADJ NO	CONT	FUNCTION	RANGE	INPUT ± 0.2 %	ADJUST		
ı	A4R4 A4R5	CHOPPER ADJ	l mV	0	AVERAGE OF ZERO Vdc @ 24		
2	A2R73	DELAY ADJ	1000 mV	1000mV	165 µs @ 29 FROM 30		
3	AIRI	FRONT PANEL ZERO	l mV	0	AVERAGE ZERO INDICATION		
4	A2R72	MASTER GAIN ADJ		_			
5	A2R27	RANGE ADJ	l mV	· I mV			
6	A2R26	RANGE ADJ	3 m∨	3 mV			
7	A2R25	RANGE ADJ	10 mV	10 mV			
8	A2R24	RANGE ADJ	30 mV	30 mV	4.00V AT (24)		
9	A2R3I	RANGE ADJ	100 mV	100 mV			
0	A2R30	RANGE ADJ	300 mV	300 mV			
11	A2R70	METER FS ADJ	1000 mV	. 1000 mV			
12	A2R28	RANGE ADJ	3000 mV	3000 mV			
13	A2R71	OUTPUT INTERGRATOR ADJ	1000 mV	1000 mV	WITH 4.00V @ 24 ADJUST de ANALOG FOR 10.00V		
14	A2R29	RANGE ADJ	1000 mV	1000 mV	+ 10.00V AT dc ANALOG TERMINALS DC VOLTMETER INPUT > 10 M OHMS		
15	A2R70	METER FS ADJ	1000 mV	1000 mV	1000 mV INDICATION		
16	A2R28	RANGE ADJ	3000 mV	3000 mV	3000 mV INDICATION		
17	A2R27	RANGE ADJ	l mV	l m∨	1.000 mV INDICATION		
18	A2R26	RANGE ADJ	3 mV	3 mV	3.000mV INDICATION		
19	A2R25	RANGE ADJ	10 mV	10 mV	10.00 mV INDICATION		
20	A2R24	RANGE ADJ	30 mV	30 mV	30.00 mV INDICATION		
21	A2R3I	RANGE ADJ	100 mV	100 mV	100.0 mV INDICATION		
22	A2R30	RANGE ADJ	300 mV	300 mV	300.0 mV INDICATION		

Figure 5-3. Adjustment Locations and Descriptions.

#### SECTION V MAINTENANCE

#### 5-25. (Continued).

- c. Connect the sensor to the adjustable signal source and set the out-put of the source to zero.
- d. Set the two chopper adjustments, A4R4 and A4R5, to their physical midpoints.
- e. Connect the precision voltmeter to TP24 and note the indication.
- f. Adjust A4R4 to decrease the voltmeter's indication to one half of that in Step d.
- q. Adjust A4R5 to bring the indication to zero. There will be some fluctuation of the indication and averaging will be required.
- $\ensuremath{\text{h}}_{\bullet}$  Re-connect the "zero" potentiometer connector, AIP1.
- **5-26.** Delay Adjustment. Preform the adjustment as follows:
- a. Adjust A2R73 so that the falling edge of the square wave at TP29 is delayed 165 ms with respect to the falling edge of the square wave at TP30.
- 5-27. Zeroing Adjustment. Perform the adjustment as follows:
- a. Turn the instrument "off" for at least one minute. If the meter pointer does not come to rest at zero, use the zero adjustment screw in the lower center of the meter to bring the pointer to zero. Turn the instrument "on"; be sure that the instrument has stabilized at room temperature at least five minutes before the following calibration is performed.
- b. Set the instrument FULL SCALE range selector to the 1 mV range and zero the instrument.
- 5-28. Master Gain Adjustments. Perform the adjustment as follows:
- a. Set A2R72 to the midpoint of its range.
- b. Set the FULL SCALE range selector to the 1 mV range, apply 1.000 mV input and adjust A2R27 for -4.00 V at TP24.
- c. Set the FULL SCALE range selector to the 3 mV range, zero the instrument, apply 3.000 mV input, and adjust A2R26 for -4.00 V at TP24.
- d. Set the FULL SCALE range selector to the 10 mV range, zero the instrument, apply 10.00 mV input, and adjust A2R25 for -4.00 V at TP24.
- e. Set the FULL SCALE range selector to the 30~mV range, apply 30.00~mV input, and adjust A2R24 for -4.00~V at TP24.
- f. Set the FULL SCALE range selector to the 100 mV range, apply 100.0 mV input, and adjust A2R31 for -4.00 V at TP24.

- g. Set the FULL SCALE range selector to the 300 mV range, apply 300.0 mV input, and adjust A2R30 for -4.00 V at TP24.
- h. Set the FULL SCALE range selector to the 1000 mV range, apply 1000 mV input, and adjust A2R29 for -4.00 V at TP24.
- i. Set the FULL SCALE range selector to the 3000 mV range, apply 3000 mV input, and adjust A2R28 for -4.00 V at TP24.
- 5-29. Output Integrator Adjustments. Perform the adjustments as follows:
- a. Set the FULL SCALE range selector to the 1000 mV range, apply 1000 mV input and check for  $-4.00~\pm.10$  V at TP24. If the reading is not within tolerance repeat the Master Gain adjustments. If the reading is within tolerance perform step b.
- b. Adjust A2R71 for 10.00 V at the DC ANALOG output.
- **5-30.** Range Adjustments. Perform the adjustments as follows:
- a. Set the FULL SCALE range selector to the 1000 mV range and apply 1000 mV input, adjust A2R29 for +10.00 V at the DC ANALOG terminals.
- b. Set the FULL SCALE range selector to the 1000 mV range and apply 1000 mV input, observe the panel meter and adjust A2R70 for 1000 mV.
- c. Set the FULL SCALE range selector to the 3000 mV range and apply 3000 mV input, observe the panel meter and adjust A2R28 for 3000 mV.
- d. Set the FULL SCALE range selector to the 1 mV range, zero the instrument, apply 1.00 mV input, observe the panel meter and adjust AZR27 for 1.00 mV.
- e. Set the FULL SCALE range selector to the 3 mV range, zero the instrument, apply 3.00 mV, observe the panel meter and adjust A2R26 for 3.00 mV.
- f. Set the FULL SCALE range selector to the 10 mV range, zero the instrument, apply 10.00 mV, observe the panel meter and adjust A2R25 for 10.00 mV.
- g. Set the FULL SCALE range selector to the 30 mV range, apply 30.00 mV input, observe the panel meter and adjust A2R24 for 30.00 mV.
- h. Set the FULL SCALE range selector to the 100 mV range, apply 100.0 mV input, observe the panel meter and adjust A2R31 for 100.0 mV.
- i. Set the FULL SCALE range selector to the 300 mV range, apply 300.0 mV input, observe the panel meter and adjust A2R30 for 300.0 mV.

#### 5-31. TROUBLESHOOTING

- 5-32. Instrument malfunction will generally be evident from front panel indications. The problems will fall into two general categories: selective failure of one subsystem or catastrophic failure.
- 5-33. Selective failure of one section of the instrument or out of specification performance will be evident from manipulation of the front panel controls. For example, incorrect or erratic indications would be evident from the display readings.
- 5-34. Catastrophic failures, on the other hand, would generally cause the instrument to be completely inoperative. For instance, if the power supply was not operating properly, the instrument would not operate.
- 5-35. Further isolation of the problem requires some understanding of the simplified block diagram. Refer to Section IV, read over the theory of operation section and then try to perform the adjustment procedures as some apparent malfunctions may be corrected by these adjustments. Failure to obtain a

correct adjustment will often help to reveal the source of trouble. If this does not correct the problem proceed with the troubleshooting section below. When the problem is localized to a specific assembly or module refer to the troubleshooting procedure as listed in Table 5-3 Troubleshooting Chart.

#### WARNING

Read the Safety Summary loacted at the front of this manual before attempting any troubleshooting procedures.

#### 5-36. TROUBLE LOCALIZATION

- 5-37. Many malfunctions are evident from the front panel meter. Other front panel indications might include erratic or incorrect meter readings. In each case the circuit most closely associated with the malfunction should be tested first.
- **5-38.** Before troubleshooting the instrument reter to paragraph 5-9, Removal and Replacement Procedures and remove the top and bottom instrument covers.

### Table 5-3. Troubleshooting Chart. **Troubleshoot** Symptom Probe Troubleshooting. instrument can not be zeroed. Refer to paragraph 4-9. No response to RF input. Incorrect response to RF input. a. Disconnect the probe from the instrument and measure the resistance between the two pins on the probe connector. The probe should read 10k to 20k ohms in one direction and infinite resistance in the other direction. b. If the probe resistance is not within tolerance, replace the probe. Refer to paragrah 5-24, perform the Adjustment Procedure and paragraph 5-17, perform the Probe Tests. If the probe resistance is within tolerance, Refer c. If the probe resistance is within tolerance, kerel to paragrah 5-24, perform the Adjustment Procedure and paragraph 5-17, perform the Probe Tests. d. If the instrument fails any part of the adjustment procedure, refer to to the specific circuitry troubleshooting procedure where the adjustment failed. Power Supply Circuitry Troubleshooting. No readings. Refer to paragraph 4-36, Figures 4-6 and 7-2. Erratic readings.

- a. Verify that the power cord is connected to the instrument and an appropriate power source.
- b. Verify that the LINE ON/OFF power switch is set to ON and the LED is lit.
- c. Verify that the line selector switches are set for the proper power.
- d. Verify that the fuse has not blown, and the proper fuse is installed.

#### SECTION VI PARTS LIST

# 6-1. INTRODUCTION

6-2. Table 6-2. Replaceable Parts, list all the replaceable parts and includes; the reference symbol, description, Mfr., Mfr's

Part No., and the BEC Part No. Table 6-1. Manufacturer's Federal Supply Code Numbers list the manufacturer's federal supply numbers.

TABLE 6-1. Manufacturer's Federal Supply Code Numbers.

NUMBER	NAME	NUMBER	NAME
01295 To 02260 Av 02735 Ri 04713 Mi 06383 Pi 06776 Ri 07263 Fi 16546 Ci 17856 Si 19701 Mi 20307 Ai 02260 Av 02007 Ai 02007 Ai 02007 Ai 02007 Ai 02007 Ai 02260 Ai 022	llen Bradley exas Instruments mphenol CA Solid State Div. otorola Semiconductor oonton Electronics anduit Corp. obinson Nugent, Inc. airchild Semiconductor entralab iliconix, Inc. epco Electra rco - Micronics olex, Inc.	27735 28480 32293 34430 51640 51791 54426 56289 57582 71450 73138 91506 98291 S4217	F-Dyne Electronics Hewlett-Packard Corp. Intersil, Inc. Monsanto Analog Devices, Inc. Statek Corp. Buss Fuses Sprague Electric Co. Kahgan CTS Corp. Beckman Instr., Helipot Div. Augat Sealectro Corp. United Chemicon, Inc.

TABLE 6-2. LIST OF REPLACEABLE PARTS.

REFERENCE DESIGNATOR	DESCRIPTION	FED CODE	MANUFACTURER PART NUMBER	отч	BEC PART NUMBER
		24224	000057004	,	09205300A
	SUB PANEL ASSEMBLY	04901	09205300A	!	09215300B
	ZERO POT ASSEMBLY, 5k	04901	09215300B	;	09205600A
	SWITCH PUSHBUTTON ASSEMBLY	04901	09205600A	į	09205700A
	SWITCH FINAL ASSEMBLY POWER	04901	09205700A	•	09207100A
	PWA POWER SWITCH	04901	09207100A	<b>;</b>	09207100A
	REAR PANEL ASSEMBLY	04901	09205201A	,	60470201A
	REAR PANEL UNIT	04901	60470201A		09215500G
	PWA MAIN	04901	09215500G	1	04216102A
	PWA CHOPPER DUROID	04901	04216102A	!	09171201A
	PROBE PWA/CONTACT/CONN ASSY	04901	09171201A	!	09171201A
	PWA PROBE	04901	09171301A		
	SUB PANEL ASSEMBLY				09205300A
		. 55, 00	7477820000	1	554216000
M1	METER AND SCALE				09205400A
W6	CABLE ASSY METER	04901	09205400A		092034008
	ZERO POT ASSEMBLY, 5k				092153008
				•	31125500A
R1	RES VAR 5k 10% 2W	04901	31125500A	1	479429000
W7P1	CONNECTOR HOUSING 4 CIR	27264	22-01-2047	'	4/9429000

TABLE 6-2. LIST OF REPLACEABLE PARTS. (CONTINUED).

REFERENCE DESIGNATOR	DESCRIPTION	FED CODE	MANUFACTURER PART NUMBER	QTY	BEC PART NUMBER
	SWITCH PUSHBUTTON ASSEMBLY				09205600A
\$2 ₩3	SWITCH PUSHBUTTON CABLE ASSY RANGE SWITCH	04901 04901		1	57117300B
	SWITCH FINAL ASSEMBLY POWER				09205700A
₩4	•	04901	09205800A		09205800A
	PWA POWER SWITCH			-	09207100A
CR1	LED YELLOW DIFF 5082-4684			1	536034000
	REAR PANEL ASSEMBLY				09205201A
F1 P2 P3 P4 S3	FUSE 0.2 AMP 250V MDL CONNECTOR 6 CIR CONNECTOR BNC CONNECTOR HOUSING SWITCH ROCKER DPDT TRANSFORMER POWER	54426 06383 54420 27264 13812 04901	MDL 0.2 CE156F22-6-C UG-625/U 22-01-2021 572-2121-0103-010 446071000	1 1 1 1	479416000 479123000 469415000 465286000
	REAR PANEL UNIT	82389	EAC309	1	60470201A 477281000
P1 S1	CONNECTOR LINE CORD SWITCH DUAL SLIDE DPDT-DPDT	82389		i	
PWA MAIN					09215500G
C4-6 C7-11 C12	CAP EL 2200µF -10\$ +50\$ 35V CAP EL 100µF 20\$ 25V CAP PE 0.1µF 10\$ 200V CAP PE 0.1µF 10\$ 100V CAP PP 0.1µF 10\$ 100V CAP PP 0.1µF 10\$ 100V CAP PP 0.1µF 10\$ 100V CAP MICA 240PF 5\$ 500V DIODE BRIDGE KBP-02 DIODE SIG 1N4001 DIODE HSCH1001 (1N6263) DIODE SIG 1N914 HEADER 6 PION STRAIGHT CONNECTOR 2 PIN STRAIGHT SOCKET SPRING COMPONENT LEAD CONNECTOR 4 PIN STRAIGHT HEADER 9 PIN STRAIGHT CONNECTOR 2 PIN STRAIGHT CONNECTOR 2 PIN STRAIGHT CONNECTOR 2 PIN STRAIGHT CONNECTOR 2 PIN STRAIGHT TONNECTOR 2 PIN STRAIGHT TRANS PNP 2N3906 TRANS NPN 2N5088 BLUE RES MF 10.0K 1\$ 1/4W RES MF 4.99K 1\$ 1/4W RES MF 2.21K 1\$ 1/4W	57582 S4217 S4217 S4217 56289 S4217 27735 16546 27735 14655 15281 04713 28480 01295 06383 27264 06383 27264 06383 27264 27264 27264 27264 27264 19701 19701	KSMM-2200-35 SM-25-VB-100-M SM-25-VB-10-M SM-25-VB-100-M 192P10492 SM-25-VB-100-M PP11-1-10-100 CE-151 PP11-1-10-100 CD15FD241J KBP02 1N4001 HSCH-1001 1N914 HPSS156-6-C 22-03-2021 1-332070-7 22-04-2021 HPSS100-9-C 22-03-2021 22-03-2021 22-03-2021 22-03-2021 22-03-2021 22-03-2021 22-03-2021 2N5088 5043ED10K00F 5043ED4K990F RN55D-2211-F	3351211 13123151 11111111111111111111111	283336000 283334000 234005000 283334000

TABLE 6-2. LIST OF REPLACEABLE PARTS. (CONTINUED).

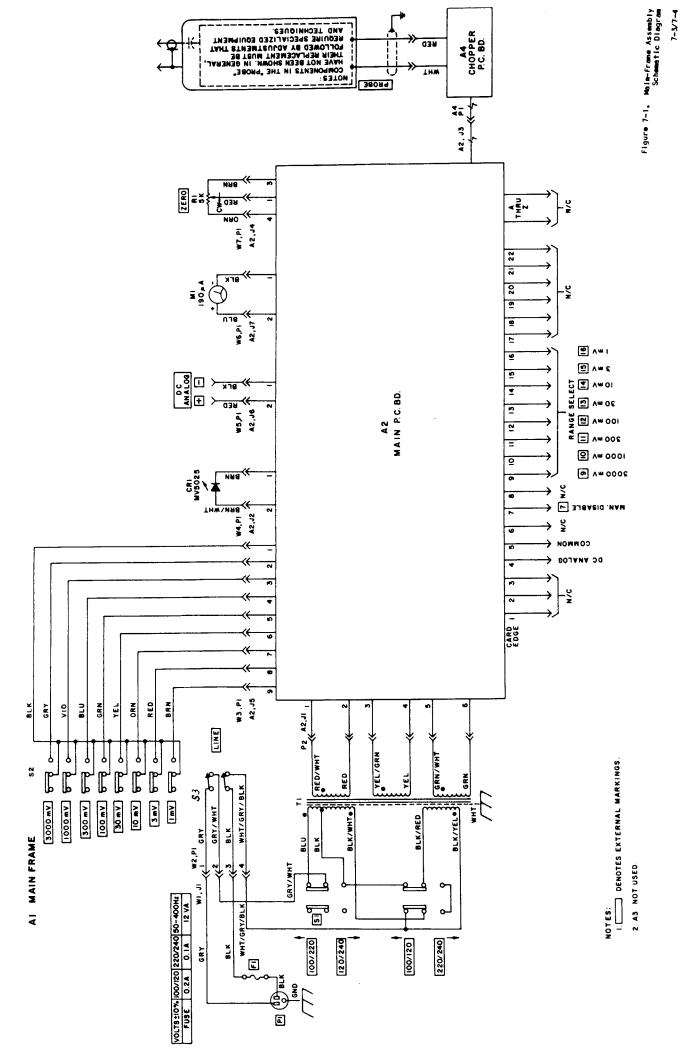
REFERENCE	DESCRIPTION	FED CODE	MANUFACTURER PART NUMBER	оту	BEC PART NUMBER
					09215500G
PWA MAIN (C	CONTINUED)				
R4	RES MF 12.7K 1% 1/4W	19701	5043ED12K70F	1	341410000 343250000
R5	RES COMP 330 OHM 5% 1/4W	01121 19701	CB3315 RN55D=9092=F	2	341492000
R6-7	RES MF 90.9K 1% 1/4W	01121	CB1265	ī	343708000
R8 R9	RES COMP 12M 5% 1/4W RES MF 165 OHM 1% 1/4W	19701	5043ED165R0F	1	341221000
R10	RES COMP 12M 5% 1/4W	01121	CB1265	1	343708000
R11	RES MF 100K 1% 1/4W	19701	5043ED100K0F	1	341500000
R12	RES MF 10.0K 1% 1/4W	19701	5043ED10K00F	1	341400000 341300000
R13	RES MF 1.00K 1% 1/4W	19701 19701	5043ED1K000F 5043ED221R0F	i	341233000
R14	RES MF 221 OHM 1% 1/4W RES MF 1.00K 1% 1/4W	19701	5043ED1K000F	į	341300000
R15 R16	RES MF 10.0K 1% 1/4W	19701	5043ED10K00F	1	341400000
R17	RES MF 100K 1% 1/4W	19701	5043ED100K0F	1	341500000
R18	RES MF 8.25K 1% 1/4W	19701	5043ED8K250F	1	341388000
R19	RES MF 1.00K 1% 1/4W	19701	5043ED1K000F	1	341300000 341388000
R20	RES MF 8.25K 1% 1/4W	19701	5043ED8K250F 5043ED1K000F	1	341300000
R21	RES MF 1.00K 1% 1/4W	19701 01121	CB1 155	i	343604000
R22	RES COMP 1.1M 5% 1/4W RES MF 10.0K 1% 1/4W	19701	5043ED10K00F	1	341400000
R23 R24 <del>-</del> 31	RES VAR 25K 10% 0.5W	73138	72PR25K	8	311400000
R32	RES MF 53.6K 1% 1/4W	19701	5043ED53K60F	1	341470000
R33	RES MF 78.7K 1% 1/4W	19701	5043ED78K70F	1	341486000
R34	RES MF 20.0K 1% 1/4W	19701	5043ED20K00F	1	341429000 341470000
R35	RES MF 53.6K 1% 1/4W	19701	5043ED53K60F 5043ED5K110F	i	341400000
R36	RES MF 5.11K 1% 1/4W	19701 19701	5043ED18K70F	i	341426000
R37 R38	RES MF 18.7K 1% 1/4W RES MF 1.00K 1% 1/4W	19701	5043ED1K000F	i	341300000
R39	RES MF 4.99K 1% 1/4W	19701	5043ED4K990F	1	341367000
R40	RES MF 590 OHM 1% 1/4W	19701	5043ED590R0F	1	341274000
R4 1	RES MF 1.33K 1% 1/4W	19701	5043ED1K330F	!	341312000
R42	RES MF 150 OHM 1% 1/4W	19701	5043ED150R0F	1	341217000 341241000
R43	RES MF 267 OHM 1% 1/4W	19701 19701	5043ED267R0F 5043ED10K00F	1	341400000
R44	RES MF 10.0K 1% 1/4W RES MF 100K 1% 1/4W	19701	5043ED100K0F	į	341500000
R45 R46	RES MF 33.2K 1% 1/4W	19701	5043ED33K20F	1	341450000
R47	RES MF 56K 1% 1/4W	01121	CB5635	1	343472000
R48-49	RES NETWORK 3.3K 2% 1.5W	71450	750-101-R3.3K	2	345030000
R50	RES MF 100K 1% 1/4W	19701	5043ED100K0F	2	341500000 341533000
R51	RES MF 221K 1% 1/4W	19701 19701	5043ED221K0F 5043ED475K0F	i	341565000
R52	RES MF 475K 1% 1/4W RES MF 221K 1% 1/4W	19701	5043ED221K0F	i	341533000
R53 R54 <b>-</b> 55	RES COMP 4.7K 5% 1/4W	01121	CB4725	2	343365000
R56	RES MF 15.0K 1% 1/4W	19701	RN550-1502-F	1	341417000
R57	RES MF 4.99K 1% 1/4W	19701	5043ED4K990F	1	341467000
R58	RES MF 511 OHM 1\$ 1/4W	19701	5043ED511R0F	1	341268000 341429000
R59	RES MF 20.0K 1% 1/4W	19701	5043ED20K00F 5043ED10K00F	1	341429000
R60	RES MF 10.0K 1% 1/4W RES MF 20.0K 1% 1/4W	19701 19701	5043ED20K00F	3	341429000
R61-63 R64	RES MF 20.0K 1% 1/4W	19701	RN55D-2212-F	ī	341433000
R65	RES MF 221K 1% 1/4W	19701	5043ED221K0F	1	341533000
R66	RES MF 10.0K 1% 1/4W	19701	5043ED10K00F	1	341400000
R67	RES MF 45.3K 1% 1/4W	19701	5043ED45K30F	1	341463000
R68	RES MF 9.09K 1% 1/4W	19701	RN55D-9091-F	) 1	341392000 341567000
R69	RES MF 499K 1% 1/4W	19701 73138	5043ED499K0F 72PR10K	i	311328000
R70 R71	RES VAR 10K 10% 0.5W RES VAR 5K 10% 0.5W	73138	72PR5K	i	311308000
R72	RES VAR 25K 10% 0.5W	73138	72PR25K	1	311400000
R73	RES VAR 1M 20% 0.5W	73138	72PR1M	1	311418000
ÜT	IC 78MGUIC VOLT REG POS	07263	uA78MGUIC	1	311418000
U2	(G) IC 79MGUIC MODIFIED	04901	53509700A	1	53509700A 53501100A
U3	IC UA7805UC WITH HARDWARE	04901 07263	53501100A uA78L05AWC	1	535044000
U4 U5	IC 78L05 VOLT REG IC 79L05 VOLT REG	04713	MC79L05ACP	i	535090000
U6	(G) IC 356B OP AMP SELECTED	04901	535062000	1	535062000

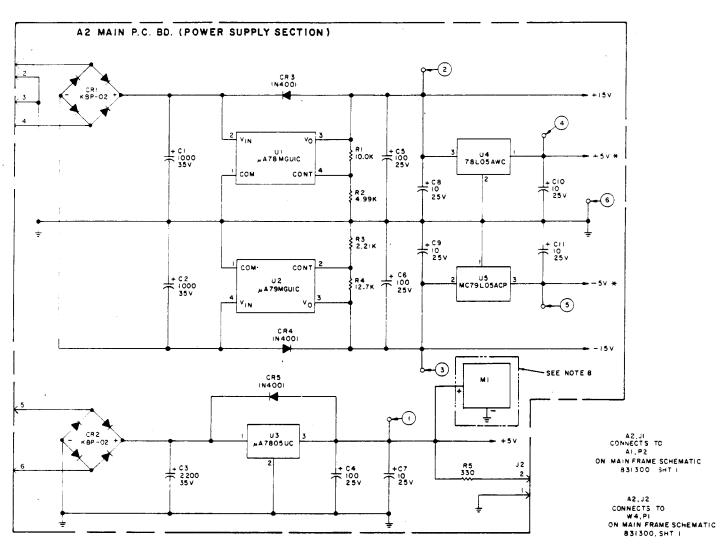
TABLE 6-2. LIST OF REPLACEABLE PARTS. (CONTINUED).

REFERENCE DESIGNATOR	DESCRIPTION	FED CODE	MANUFACTURER PART NUMBER	QTY	BEC PART NUMBER
PWA MAIN (C	CONTINUED)		IH6208CPE 535062000 CD4001AE TL074CN IH6108CPE CD4066AE CD4081BE CD4013BE CD4012BE CD4012BE SN74LS244N CD4011AE CD4040BE CD4556BE 53439100B TL43CLP DG211CJ AD7523JN TL074CN DG211CJ CD4030AE CA3140AE CD4013BE CD4013BE CD4013BE CD4013BE CD4056BE		09215500G
U7	IC 6208 4 CHAN DIF MULTPXR	32293	1H6208CPE	1	535426600
U8	(G) IC 356B OP AMP SELECTED	04901	535062000	1	535062000
U9	IC 4001A QUAD 2 INPUT NOR	02735	CD4001AE	1	534023000
U10	IC TLO74CN OP AMP QUAD	01295	TL074CN	1	535082000
U11	IC 6108 CMOS CHAN MULTPXR	32293	1H6 108CPE	1	534265000 534078000
U12	IC 4066A CMOS BILAT SW	02735	CD4000AE	1	534142000
U13	IC 4081B 2 INPUT NAND	02735 02735	CD4001BE	1	534205000
U14 U15	IC 4013B DUAL FLIP FLOP IC 4012B CMOS NAND GATE	02735	CD4013BE	i	534325000
U16	1C 4012B CMOS NAND GATE	02735	CD4013BF	i	534205000
U17	IC 4013B CMOS NAND GATE	02735	CD4012BE	1	534325000
U18	IC 741 S244 OCT BUFF & LINE DRVR	01295	SN74LS244N	1	534247000
U19	IC 74LS244 OCT BUFF & LINE DRVR IC 4011 QUAD 2 INPUT NAND	02735	CD4011AE	1	534022000
U20	IC ADADO COUNTED/DIVIDED	02735	CD4040BE	1	534275000
U21	1C 4556B DUAL 1 OF 4 DECODER	02735	CD4556BE	1	534324000
U22	IC EPROM PROG. 391AB 92E	04901	53439100B	1	53439100B
U23	IC TL431CPC ADJ PREC SHUNT REG IC 211 QUAD SPST CMOS SWITCH	01295	TL43CLP	1	534098000
U24	IC 211 QUAD SPST CMOS SWITCH	17856	DG211CJ	]	534326000
U25	10 7523 D/A CONVERTER	21040	AD7523JN	1	534099000
U26	IC TLO74CN OP AMP QUAD	01295	TE074CN	!	535082000 534326000
U27	IC 211 QUAD SPST CMOS SWITCH	17856	0040704F	1	534087000
U28	IC 4030A EXCLUSIVE OR	02735 02735	C04030AE		534050000
U29 U30	IC 3140 OP AMP	02735	CD4013RE	1	534205000
U31	IC 40138 DUAL PLIP PLOP	02735	CD4071BF	i	534141000
XU6	SOCKET IC R PIN	06776	ICN-083-S3-G	i	473041000
XU7	SOCKET IC 16 PIN	06776	ICN-163-S3-G	1	473042000
XU8	SOCKET IC 8 PIN	06776	ICN-083-S3-G	1	473041000
XU10	SOCKET IC 14 PIN	06776	ICN-143-S3-G	1	473019000
XU11	SOCKET IC 16 PIN	06776	ICN-163-S3-G	1	473042000
XU15	SOCKET IC 14 PIN	06776	ICN-143-S3-G	1	473019000
XU17	SOCKET IC 14 PIN	06776	1CN-143-S3-G	1	473019000
XU18	SOCKET IC 20 PIN	06776	ICN-203-S3-G	1	473065000
XU20	SOCKET IC 16 PIN	06776	ICN-163-S3-G	1	473042000
XU22	SOCKET IC 24 PIN	06776	1CN-246-S3-G	1	473043000 473042000
XU24	SOCKET IC 16 PIN	06776	1CN-163-S3-G 1CN-143-S3-G	i	473019000
XU26 XU27	COCKET IC 14 PIN	06776 06776	1CN-163-S3-G	i	473042000
Y1	CRYSTAL OF OU KHT	51791	CX1H96.0-C	i	547036000
	IC 3140 OP AMP IC 4013B DUAL FLIP FLOP IC 4071B CMOS 2 INPUT OR SOCKET IC 8 PIN SOCKET IC 16 PIN SOCKET IC 14 PIN SOCKET IC 14 PIN SOCKET IC 14 PIN SOCKET IC 14 PIN SOCKET IC 20 PIN SOCKET IC 16 PIN	J.,,,			
PWA CHOPPER	DUROID				04216102A
C1-2	CAP PP 0.1µF 10% 100V	27735	PP111-10-100	2	
1C1	(G) IC SELECTED QUAD SWITCH	04901	534223000	1	
P1	TERMINAL .040 OD .270 LG .062M	98291	229-1071-230	1	
P2	CONNECTOR PIN	71279	460-1521-02-03-00 5043ED51K10F	1	477400000
R3	RES MF 51.1K 1\$ 1/4W	19701		1	341468000 311400000
R4-5	RES VAR 25K 10% 0.5W	73138 19701	72PR25K 5043ED51K10F		341468000
R6 XIC1	RES MF 51.1K 1% 1/4W SOCKET IC 14 PIN	91506	714AG1D	i	
	ONTACT/CONN ASSY				09171201A
C1	CAP CER 1400pF 20% 100V	55153	D01BU142M1PX	1	224126000
CR1-2	DIODE SELECTED IT/F530022	04901	530181000	2	

TABLE 6-2. LIST OF REPLACEABLE PARTS. (CONTINUED).

REFERENCE DESIGNATOR	DESCRIPTION	FED CODE	MANUFACTURER PART NUMBER	QTY	BEC PART NUMBER
PWA PROBE					09171301A
C2-3	CAP CER CHIP 1000pF 10% 50V	61637	C1210C102K5XAH	2	224286000
R1-2(SEL)	RES CHIP 10 OHM 5% 1/8W	01121	BCD10ROJ	2	33990001A
R1-2(SEL)	RES CHIP 24 OHM 5% 1/8W	01121	BCD24ROJ	2	33990002A
R1-2(SEL)	RES CHIP 33 OHM 5% 1/8W	01121	BCD33ROJ	2	33990003A
R1-2(SEL)	RES CHIP 51 OHM 5% 1/8W	01121	BCD5 1ROJ	2	33990004A
R1-2(SEL)	RES CHIP 56 OHM 5% 1/8W	01121	BCD56ROJ	2	33990005A
R1-2(SEL)	RES CHIP 62 OHM 5% 1/8W	01121	BCD62ROJ	2	33990006A
R1-2(SEL)	RES CHIP 68 OHM 5% 1/8W	01121	BCD68ROJ	2	33990007A
R1-2(SEL)	RES CHIP 75 OHM 5% 1/8W	01121	BCD75ROJ	2	33990008A
R1-2(SEL)	RES CHIP 82 OHM 5% 1/8W	01121	BCD82ROJ	2	33990009A
	RES CHIP 91 OHM 5% 1/8W	01121	BCD91ROJ	2	33990010A
R1-2(SEL) R3-4	RES CHIP 330 OHM 5% 1/8W	01121	BCD3300J	2	33990011A





NOTES

- TOTIES

  I CAPACITANCE VALUES IN AF UNLESS OTHERWISE SPECIFIED.

  2 RESISTANCE VALUES IN OHMS

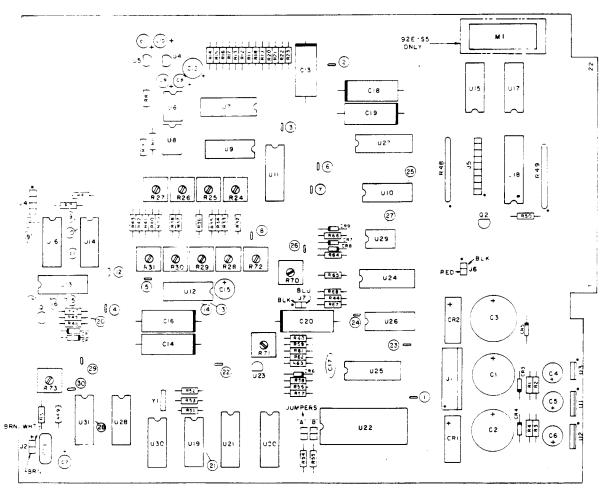
  3 SELECTED VALUES

  4 ANALOG GROUND.

- JUMPERS "A" B"B" DETERMINED BY SENSOR CHARACTERISTICS.
- 6 LAST NUMBERS USED:
- U31, R73 C21, CRII 7 TEST CONDITIONS: 1 VRMS AT INPUT TO SENSOR.
- 8 USED ON 92E-S5 ONLY.

SCHEMATIC MAIN P. C. BOARD D831300 D , SHT. 2a of 2

Figure 7-2. Main PC Board (Power Supply Section) Schematic Diagram



831301E

Figure 7-3. Main PC Board Parts Location Diagram

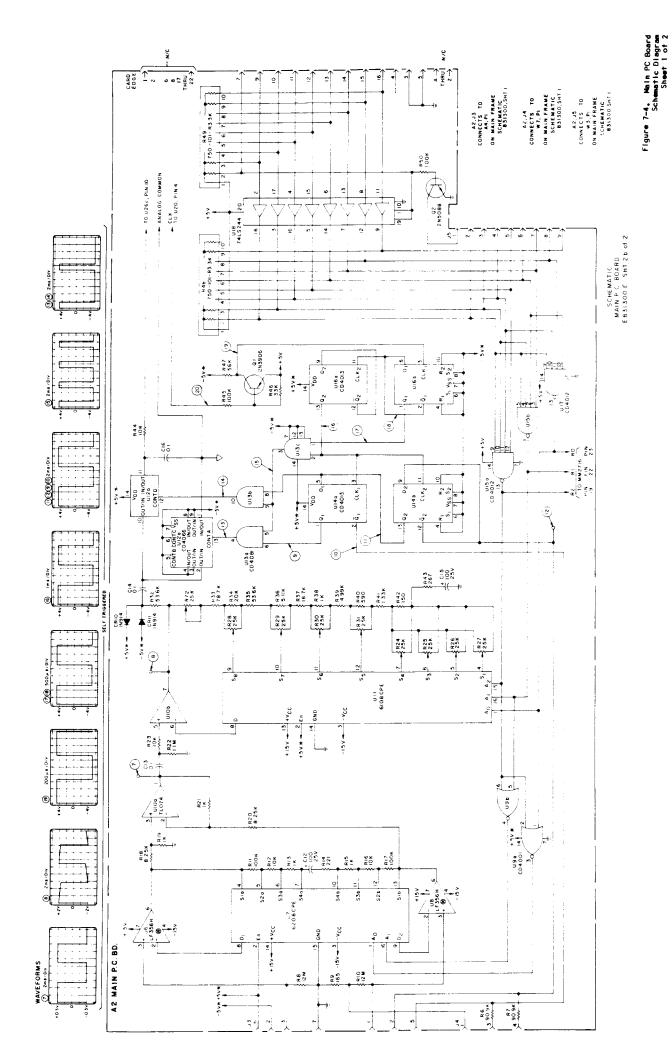


Figure 7-5, Mein PC Board Schematic Diagram Sheet 2 of 2

